Outline:

- when does a dynamical system have a unique solution?
- state-space and transfer-function representations of LTI systems
- state-space canonical forms
- matrix computation: determinants

1 State-space and transfer-function descriptions of LTI systems

- why are we learning them?
- relations between the two:

Table 1: Relations between state-space (ss) and transfer-function (tf) system representations

SS	tf
$\dot{x}(t) = Ax(t) + Bu(t), \ y(t) = Cx(t) + Du(t)$	$G(s) = C(sI - A)^{-1}B + D$
x(k+1) = Ax(k) + Bu(k), y(k) = Cx(k) + Du(k)	$G(z) = C(zI - A)^{-1}B + D$

2 Math review

2.1 Computing determinants

• 2×2 matrices:

$$\det \left[\begin{array}{cc} a & b \\ c & d \end{array} \right] = ad - bc$$

• 3×3 matrices:

$$\det \begin{bmatrix} a & b & c \\ d & e & f \\ g & h & k \end{bmatrix} = a \det \begin{bmatrix} e & f \\ h & k \end{bmatrix} - b \det \begin{bmatrix} d & f \\ g & k \end{bmatrix} + c \det \begin{bmatrix} d & e \\ g & h \end{bmatrix}$$
$$= aek + bfg + cdh - gec - bdk - ahf$$

2.2 Computing the inverse of a matrix

There are several ways to compute a matrix inverse. One approach for low-order matrices is the method of using adjugate matrix (aka adjoint matrix):

$$A^{-1} = \frac{1}{\det\left(A\right)} adj\left(A\right)$$

We explain the computation by two examples:

• 2×2 example:

$$\begin{bmatrix} a & b \\ c & d \end{bmatrix}^{-1} = \frac{1}{ad - bc} \begin{bmatrix} (-1)^{1+1}d & (-1)^{1+2}b \\ (-1)^{2+1}c & (-1)^{2+2}a \end{bmatrix}$$

where b in $(-1)^{1+2} b$ is obtained by:

- noticing b is at row 1 column 2 of A;
- looking at the element at row 2 column 1 of A;
- constructing a submatrix of A by removing row 2 and column 1 from it, i.e., [b] in this 2×2 example;
- computing the determinant of this submatrix.

- adding $(-1)^{1+2}$ as a scalar

• 3×3 example:

where $|\cdot|$ denotes the determinant of a matrix. Similar as before, the row 1 column 2 element $-\begin{vmatrix} b & c \\ h & k \end{vmatrix}$ is obtained via

$$(-1)^{2+1} \det \left(A \text{ with } [d, e, f], \begin{bmatrix} a \\ d \\ g \end{bmatrix} \text{ removed} \right)$$

3 Canonical forms

$$G = \frac{b_2 s^2 + b_1 s + b_0}{s^3 + a_2 s^2 + a_1 s + a_0}$$

• Controllable canonical form

$$A = \begin{bmatrix} 1 & \\ & 1 \\ -a_0 & -a_1 & -a_2 \end{bmatrix}, B = \begin{bmatrix} 0 \\ 0 \\ 1 \end{bmatrix}, C = \begin{bmatrix} b_0 & b_1 & b_2 \end{bmatrix}, D = 0$$

- understanding the formula: the transfer function is given by $G(s) = C(sI - A)^{-1}B + D$, where the poles of the system come from det (sI - A) = 0. Suppose we don't know the order of a_0 , a_1 and a_2 in the last row of A, and use \star as a temporary representation in A. Looking at

$$\det (sI - A) = \det \begin{bmatrix} s & -1 \\ s & -1 \\ \star & \star & s + \star \end{bmatrix}$$

we see that the only way for s^2 to appear is from the term $s^2 (s + \star)$ in the determinant computation. Hence the location of $-a_2$ has to be at the bottom right corner of A.

- exercise: write down the controllable canonical form for the following systems

*
$$G(s) = \frac{s^2 + 1}{s^3 + 2s + 10}$$

* $G(s) = \frac{b_0 s^2 + b_1 s + b_2}{s^3 + a_0 s^2 + a_1 s + a_2}$

• Observable canonical form

$$A = \begin{bmatrix} -a_2 & 1 \\ -a_1 & 1 \\ -a_0 & -a_0 \end{bmatrix}, B = \begin{bmatrix} b_2 \\ b_1 \\ b_0 \end{bmatrix}, C = \begin{bmatrix} 1 & 0 & 0 \end{bmatrix}, D = 0$$

• Diagonal form: for systems described by

$$G(s) = \frac{k_1}{s - p_1} + \frac{k_2}{s - p_2} + \frac{k_3}{s - p_3}$$

one state-space form is

$$A = \begin{bmatrix} p_1 & & \\ & p_2 & \\ & & p_3 \end{bmatrix}, B = \begin{bmatrix} 1 \\ 1 \\ 1 \end{bmatrix}, C = \begin{bmatrix} k_1 & k_2 & k_3 \end{bmatrix}, D = 0$$

• Jordan form: if

$$G(s) = \frac{k_1}{s - p_1} + \frac{k_2}{(s - p_m)^2} + \frac{k_3}{s - p_m}$$

then one state-space form of the system is

$$A = \begin{bmatrix} p_1 & & \\ & p_m & 1 \\ & & p_m \end{bmatrix}, B = \begin{bmatrix} 1 \\ 0 \\ 1 \end{bmatrix}, C = \begin{bmatrix} k_1 & k_2 & k_3 \end{bmatrix}, D = 0$$

– understanding the formula: why $B = [1, 0, 1]^T$? What if

$$G(s) = \frac{k_1}{(s-p_1)^2} + \frac{k_1}{s-p_1} + \frac{k_2}{(s-p_m)^2} + \frac{k_3}{s-p_m}$$

• Modified Jordan form: this is for systems with complex poles:

$$G(s) = \frac{k_1}{s - p_1} + \frac{\alpha s + \beta}{(s - \sigma)^2 + \omega^2}$$

we have

$$A = \begin{bmatrix} p_1 & & \\ & \sigma & \omega \\ & -\omega & \sigma \end{bmatrix}, B = \begin{bmatrix} 1 \\ 0 \\ 1 \end{bmatrix}, C = \begin{bmatrix} k_1 & k_2 & k_3 \end{bmatrix}, D = 0$$