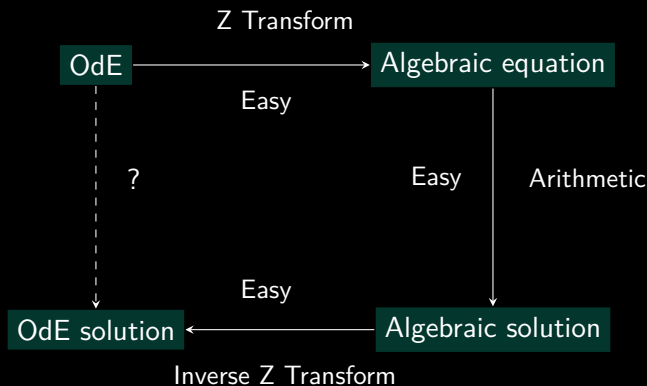


# ME547: Linear Systems Modeling: Z Transform

Xu Chen

University of Washington

# The Z transform approach to Ordinary difference Equations (OdEs)



- ▶ analogous to Laplace transform for continuous-time signals

# Definition

- ▶ let  $x(k)$  be a real discrete-time sequence that is zero if  $k < 0$
- ▶ the (one-sided) Z transform of  $x(k)$  is

$$\begin{aligned} X(z) &\triangleq \mathcal{Z}\{x(k)\} = \sum_{k=0}^{\infty} x(k)z^{-k} \\ &= x(0) + x(1)z^{-1} + x(2)z^{-2} + \dots \end{aligned}$$

where  $z \in \mathbb{C}$

- ▶ a linear operator:  $\mathcal{Z}\{\alpha f(k) + \beta g(k)\} = \alpha \mathcal{Z}\{f(k)\} + \beta \mathcal{Z}\{g(k)\}$
- ▶ the series  $1 + \gamma + \gamma^2 + \dots$  converges to  $\frac{1}{1-\gamma}$  for  $|\gamma| < 1$  [region of convergence (ROC)]
- ▶ (also, recall that  $\sum_{k=0}^N \gamma^k = \frac{1-\gamma^{N+1}}{1-\gamma}$  if  $\gamma \neq 1$ )

# Example: geometric sequence $\{a^k\}_{k=0}^{\infty}$

$$\sum_{k=0}^{\infty} \gamma^k = \frac{1}{1-\gamma}$$

▶  $x(k) = a^k$

▶  $\mathcal{Z}\{a^k\} = \sum_{k=0}^{\infty} a^k z^{-k} = \frac{1}{1 - az^{-1}} = \frac{z}{z-a}$

## Example: step sequence (discrete-time unit step function)

$$\mathcal{Z}\{a^k\} = \frac{1}{1 - az^{-1}}$$

$$\blacktriangleright 1(k) = \begin{cases} 1, & \forall k = 1, 2, \dots \\ 0, & \forall k = \dots, -1, 0 \end{cases}$$

$$\blacktriangleright \mathcal{Z}\{1(k)\} = \mathcal{Z}\{a^k\}|_{a=1} = \frac{1}{1 - z^{-1}} = \frac{z}{z-1}$$

## Example: discrete-time impulse

$$\begin{aligned} \blacktriangleright \delta(k) &= \begin{cases} 1, & k = 0 \\ 0, & \text{otherwise} \end{cases} \\ \blacktriangleright \mathcal{Z}\{\delta(k)\} &= 1 \end{aligned}$$

Exercise:  $\cos(\omega_0 k)$

$f(k)$	$F(z)$	ROC
$\delta(k)$	1	All $z$
$a^k 1(k)$	$\frac{1}{1 - az^{-1}}$	$ z  >  a $
$-a^k 1(-k - 1)$	$\frac{1}{1 - az^{-1}}$	$ z  <  a $
$ka^k 1(k)$	$\frac{az^{-1}}{(1 - az^{-1})^2}$	$ z  >  a $
$-ka^k 1(-k - 1)$	$\frac{az^{-1}}{(1 - az^{-1})^2}$	$ z  <  a $
$\cos(\omega_0 k)$	$\frac{1 - z^{-1} \cos(\omega_0)}{1 - 2z^{-1} \cos(\omega_0) + z^{-2}}$	$ z  > 1$
$\sin(\omega_0 k)$	$\frac{z^{-1} \sin(\omega_0)}{1 - 2z^{-1} \cos(\omega_0) + z^{-2}}$	$ z  > 1$
$a^k \cos(\omega_0 k)$	$\frac{1 - az^{-1} \cos(\omega_0)}{1 - 2az^{-1} \cos(\omega_0) + a^2 z^{-2}}$	$ z  >  a $
$a^k \sin(\omega_0 k)$	$\frac{az^{-1} \sin(\omega_0)}{1 - 2az^{-1} \cos(\omega_0) + a^2 z^{-2}}$	$ z  >  a $



## Properties of Z transform: time shift

- ▶ let  $\mathcal{Z}\{x(k)\} = X(z)$  and  $x(k) = 0 \forall k < 0$
- ▶ one-step delay:

$$\begin{aligned}\mathcal{Z}\{x(k-1)\} &= \sum_{k=0}^{\infty} x(k-1)z^{-k} = \sum_{k=1}^{\infty} x(k-1)z^{-k} + x(-1) \\ &= \sum_{k=1}^{\infty} x(k-1)z^{-(k-1)}z^{-1} + x(-1) \\ &= z^{-1}X(z) + x(-1) = \boxed{z^{-1}X(z)}\end{aligned}$$

- ▶ analogously,  $\mathcal{Z}\{x(k+1)\} = \sum_{k=0}^{\infty} x(k+1)z^{-k} = \boxed{zX(z) - zx(0)}$
- ▶ thus, if  $x(k+1) = Ax(k) + Bu(k)$  and  $x(0) = 0$ ,

$$zX(z) = AX(z) + BU(z) \Rightarrow X(z) = (zI - A)^{-1}BU(z)$$

provided that  $(zI - A)$  is invertible

# Solving difference equations

Solve the difference equation

$$y(k) + 3y(k-1) + 2y(k-2) = u(k-2)$$

where  $y(-2) = y(-1) = 0$  and  $u(k) = 1(k)$ .

- ▶  $\mathcal{Z}\{y(k-1)\} = z^{-1}\mathcal{Z}\{y(k)\} = z^{-1}Y(z)$
- ▶  $\mathcal{Z}\{y(k-2)\} = z^{-1}\mathcal{Z}\{y(k-1)\} = z^{-2}Y(z)$
- ▶  $\mathcal{Z}\{U(k-2)\} = z^{-2}U(z)$
- ▶  $\Rightarrow (1 + 3z^{-1} + 2z^{-2})Y(z) = z^{-2}U(z)$
- ▶  $\Rightarrow \boxed{Y(z) = \frac{1}{z^2 + 3z + 2}U(z)}$

# Solving difference equations

Solve the difference equation

$$y(k) + 3y(k-1) + 2y(k-2) = u(k-2)$$

where  $y(-2) = y(-1) = 0$  and  $u(k) = 1(k)$ .

- ▶ 
$$Y(z) = \frac{1}{z^2 + 3z + 2} U(z) = \frac{1}{(z+2)(z+1)} U(z)$$
- ▶  $u(k) = 1(k) \Rightarrow U(z) = 1/(1 - z^{-1})$
- ▶  $\Rightarrow Y(z) = \frac{z}{(z-1)(z+2)(z+1)} = \frac{1}{6} \frac{z}{z-1} + \frac{1}{3} \frac{z}{z+2} - \frac{1}{2} \frac{z}{z+1}$  (careful with the partial fraction expansion)
- ▶ inverse Z transform then gives  
$$y(k) = \frac{1}{6} 1(k) + \frac{1}{3} (-2)^k - \frac{1}{2} (-1)^k, \quad k \geq 0$$

# From difference equation to transfer functions

- ▶ general discrete-time OdE:

$$y(k) + a_{n-1}y(k-1) + \cdots + a_0y(k-n) = b_m u(k+m-n) + \cdots + b_0 u(k-n)$$

where  $y(k) = 0 \quad \forall k < 0$

- ▶ applying Z transform to the OdE yields

$$(z^n + a_{n-1}z^{n-1} + \cdots + a_0) Y(z) = (b_m z^m + b_{m-1}z^{m-1} + \cdots + b_0) U(z)$$

- ▶ hence

$$Y(z) = \underbrace{\frac{b_m z^m + b_{m-1}z^{m-1} \cdots + b_1 z + b_0}{z^n + a_{n-1}z^{n-1} + \cdots + a_1 z + a_0}}_{G_{yu}(z): \text{ discrete-time transfer function}} U(z)$$

# DC gain of discrete-time transfer functions

- ▶ general discrete-time OdE and transfer function:

$$y(k) + a_{n-1}y(k-1) + \cdots + a_0y(k-n) = b_m u(k+m-n) + \cdots + b_0u(k-n)$$

$$Y(z) = \underbrace{\frac{b_m z^m + b_{m-1} z^{m-1} \cdots + b_1 z + b_0}{z^n + a_{n-1} z^{n-1} + \cdots + a_1 z + a_0}}_{G_{yu}(z): \text{ discrete-time transfer function}} U(z)$$

- ▶ assuming constant input and convergent output, then at steady state,

- ▶  $y(k) = y(k-1) = \cdots = y(k-n) \triangleq y_{ss}$  and  
 $u(k+m-n) = u(k+m-n-1) = \cdots = u(k-n) \triangleq u_{ss}$
- ▶  $y_{ss} + a_{n-1}y_{ss} + \cdots + a_0y_{ss} = b_mu_{ss} + \cdots + b_0u_{ss}$

- ▶ thus,

$$\underline{\text{DC gain of } G_{yu}(z)} = \frac{b_m + b_{m-1} + \cdots + b_0}{1 + a_{n-1} + \cdots + a_0} = \underline{G_{yu}(z)} \Big|_{z=1}$$

# Transfer functions in two domains

$$y(k) + a_{n-1}y(k-1) + \dots + a_0y(k-n) = b_m u(k+m-n) + \dots + b_0 u(k-n)$$
$$\iff G_{yu}(z) = \frac{B(z)}{A(z)} = \frac{b_m z^m + b_{m-1} z^{m-1} \dots + b_1 z + b_0}{z^n + a_{n-1} z^{n-1} + \dots + a_1 z + a_0}$$

v.s.

$$\frac{d^n y(t)}{dt^n} + a_{n-1} \frac{d^{n-1} y(t)}{dt^{n-1}} + \dots + a_0 y(t) = b_m \frac{d^m u(t)}{dt^m} + b_{m-1} \frac{d^{m-1} u(t)}{dt^{m-1}} + \dots + b_0 u(t)$$
$$\iff G_{yu}(s) = \frac{B(s)}{A(s)} = \frac{b_m s^m + \dots + b_1 s + b_0}{s^n + a_{n-1} s^{n-1} + \dots + a_1 s + a_0}$$

Properties	$G_{yu}(s)$	$G_{yu}(z)$
poles and zeros	roots of $A(s)$ and $B(s)$	roots of $A(z)$ and $B(z)$
causality condition	$n \geq m$	$n \geq m$
DC gain / steady-state response to unit step	$G_{yu}(0)$	$G_{yu}(1)$

## Additional useful properties of Z transform

- ▶ time shifting (assuming  $x(k) = 0$  if  $k < 0$ ):

$$\mathcal{Z} \{x(k - n_d)\} = z^{-n_d} X(z)$$

- ▶ Z-domain scaling:  $\mathcal{Z} \{a^k x(k)\} = X(a^{-1}z)$

- ▶ differentiation:  $\mathcal{Z} \{kx(k)\} = -z \frac{dX(z)}{dz}$

- ▶ time reversal:  $\mathcal{Z} \{x(-k)\} = X(z^{-1})$

- ▶ convolution: let  $f(k) * g(k) \triangleq \sum_{j=0}^k f(k-j)g(j)$ , then

$$\mathcal{Z} \{f(k) * g(k)\} = F(z) G(z)$$

- ▶ initial value theorem:  $f(0) = \lim_{z \rightarrow \infty} F(z)$

- ▶ final value theorem:  $\lim_{k \rightarrow \infty} f(k) = \lim_{z \rightarrow 1} (z-1)F(z)$ , if  $\lim_{k \rightarrow \infty} f(k)$  exists and is finite

# Mortgage payment

- ▶ imagine you borrow \$100,000 (e.g., for a mortgage)
- ▶ annual percent rate:  $APR = 4.0\%$
- ▶ plan to pay off in 30 years with fixed monthly payments
- ▶ interest computed monthly
- ▶ what is your monthly payment?



# Mortgage payment

- ▶ borrow \$100,000  $\Rightarrow$  initial debt  $y(0) = 100,000$
- ▶  $APR = 4.0\% \Rightarrow MPR = \frac{4.0\%}{12} = 0.0033$
- ▶ pay off in 30 years ( $N = 30 \times 12 = 360$  months)  $\Rightarrow y(N) = 0$
- ▶ debt at month  $k + 1$ :

$$y(k+1) = \underbrace{(1 + MPR)}_a y(k) - \underbrace{b}_{\text{monthly payment}} 1(k)$$

- ▶  $\Rightarrow Y(z) = \frac{z}{z-a}y(0) + \frac{1}{z-a} \frac{b}{1-z^{-1}}$
- ▶  $\Rightarrow Y(z) = \frac{1}{1-az^{-1}}y(0) + \frac{b}{1-a} \left( \frac{1}{1-az^{-1}} - \frac{1}{1-z^{-1}} \right)$
- ▶  $\Rightarrow y(k) = a^k y(0) + \frac{b}{1-a} (a^k - 1)$
- ▶ need  $y(N) = 0 \Rightarrow a^N y(0) = -\frac{b}{1-a} (a^N - 1)$
- ▶  $\Rightarrow b = \frac{a^N y(0)(a-1)}{a^N - 1} = \$477.42$